

# Spatial Decomposition-Based Fault Detection Framework for Parabolic-Distributed Parameter Processes

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**Abstract**—Fault detection for distributed parameter processes (heat processes, fluid processes, etc.) is vital for safe and efficient operation. On one hand, the existing data-driven methods neglect the evolution dynamics of the processes and cannot guarantee that they work for highly dynamic or transient processes; on the other hand, model-based methods reported so far are mostly based on the backstepping technique, which does not possess enough redundancy for fault detection since only the boundary measurement is considered. Motivated by these considerations, we intend to investigate the robust fault detection problem for distributed parameter processes in a model-based perspective covering both boundary and in-domain measurement cases. A real-time fault detection filter (FDF) is presented, which gets rid of a large amount of data collection and offline training procedures. Rigorous theoretic analysis is presented for guiding the parameters selection and threshold computation. A time-varying threshold is designed such that the false alarm in the transient stage can be avoided. Successful application results on a hot strip mill cooling system demonstrate the potential for real industrial applications.

**Index Terms**—Distributed parameter process, fault detection, partial differential equation (PDE) observer.

## I. INTRODUCTION

**D**ISTRIBUTED parameter processes widely exist in industrial applications, for example, transport-reaction processes [1], snap curing processes [2]–[4], and battery thermal processes [5]–[7]. Generally, these processes can

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all be described by partial differential equations (PDEs) mathematically, which possess the spatial-temporal dynamics [8]–[12]. With the rapidly increasing demand for the high-production quality with economic operations for these industrial processes, the requirement of system safety is a critical issue that needs to be addressed. Faults hidden in these processes may cause system failure or even permanent damage if they are not discovered in time. For example, the distributed thermal fault in lithium-ion (Li-ion) batteries [5], [13] can lead to battery degradation, failure, and even thermal runaway. The in-domain actuator faults of diffusion-reaction processes [1], [14] that widely exist in chemical processes can cause severe effects. Despite the fact that advanced control design [15], [16] for distributed parameter systems (DPSs) (including sampled-data [17], [18] and fuzzy control [19]) were well studied in recent years, works reported on fault diagnosis for distributed parameter processes are relatively rare so far.

On one hand, for decades, the fault detection and fault-tolerant control for technical processes that are modeled by ordinary differential equations (ODEs) have been investigated sufficiently, from both model-based [20]–[22] and data-driven [23] perspectives; on the other hand, research focusing on the fault diagnosis of PDEs is relatively rare. Similarly, these works can be roughly separated into these two subcategories.

In terms of the data-driven fault detection for distributed parameter processes, some works have been reported recently (see [24]–[27] for reference). For example, by using the Galerkin method, a data-driven process monitoring approach for parabolic DPS was discussed [24]. A novel data-driven method was proposed for fault detection and localization of parabolic DPSs [25]. However, these methods either used modal approximation [24] which induces *observation spillover* or neglected the evolution dynamics of the processes [25], [27] and cannot guarantee work for highly dynamic or transient processes.

As for the existing model-based fault detection methods, they can be further divided into two subcategories: 1) the so-called “early-lumping” [28] approach that reduces the PDE into several approximated ODEs in the first place and then analyzes the ODE model (representative works can be referred to [1], [14], and [29]). The other was the so-called “late-lumping” approach [30], [31] which directly investigates the PDE system without approximation. Research belong to this category is mainly based on the backstepping

design techniques [32] (see [13] and [33]). “Early-lumping”-based approaches have been criticized for the *observation spillover* induced by the modal approximation used. Moreover, for the backstepping-based fault detection methods, the redundancy cannot be guaranteed for certain applications since only boundary measurement is used. Hence, a unified framework for fault detection, which can be applied to either the boundary and in-domain sensor, is still missing.

Considering the above facts, we are devoted to the study of a methodology framework for robust fault detection of distributed parameter processes from a model-based perspective while covering both boundary and in-domain measurement cases. To be more specific, a fault detection filter (FDF) is first constructed to generate signals by mimicking the FDF design of lumped parameter systems (LPSs) [20]. Furthermore, the residual evaluation scheme and threshold computation procedure are conducted in a similar manner.

The remainder of this article can be summarized as follows. In Section II, the detailed problem statement is introduced. The methodology framework is illustrated in Section III. The Luenberger-type PDE observer and FDF design are presented in Sections IV and V, respectively. Illustrative demonstrations on a hot strip mill cooling system are demonstrated in Section VI. Finally, the concluding remarks are presented in Section VII.

## II. PRELIMINARIES AND PROBLEM STATEMENT

*Notations:*  $\mathbb{R}^n$  denotes  $n$ -dimensional Euclidean space, and the norm is denoted as  $\|\cdot\|_2$ .  $\mathcal{L}^2([0, L]) \triangleq \mathcal{L}^2((0, L); \mathbb{R})$  is a real Hilbert space of square integrable functions  $\omega(x) : [0, L] \rightarrow \mathbb{R}$ , and the spatial  $\mathcal{L}^2$  norm is defined by  $\|\omega(\cdot)\|_2 \triangleq \sqrt{\int_0^L \omega^2(x) dx}$ . Given a natural number  $\bar{l}$ ,  $\mathcal{H}^{\bar{l}}(0, 1) \triangleq \mathcal{W}^{\bar{l}, 2}((0, 1); \mathbb{R})$  is a real Sobolev space of absolutely continuous functions  $\bar{\omega}(x) : (0, 1) \rightarrow \mathbb{R}$  with square integrable derivatives  $d^i \bar{\omega}(x)/dx^i$  up to the order  $\bar{l} \geq 1$  and with the norm  $\|\bar{\omega}(\cdot)\|_{\mathcal{H}^{\bar{l}}} \triangleq \sqrt{\int_0^1 \sum_{i=0}^{\bar{l}} (d^i \bar{\omega}(x)/dx^i)^2 dx}$ . For any functions  $\omega(\cdot) \in \mathcal{H}^1(0, 1)$ , the spatial  $\mathcal{L}^\infty$  norm is defined as  $\|\omega(\cdot)\|_\infty \triangleq \max_{x \in [0, 1]} |\omega(x)|$ .  $T_t \triangleq \partial T / \partial t$ ,  $T_x \triangleq \partial T / \partial x$ ,  $T_{xx} \triangleq \partial^2 T / \partial x^2$ .

### A. Process Description

We consider the distributed parameter processes that can be described by the following PDE:

$$T_t(x, t) = T_{xx}(x, t) + d(x, t) + f(x, t) \quad (1a)$$

$$T_x(0, t) = \eta_1 T(0, t) \quad (1b)$$

$$T_x(1, t) = -\eta_2 T(1, t) \quad (1c)$$

$$T(x, 0) = T_0(x) \quad (1d)$$

$$\mathbf{y}(t) = \int_0^1 \mathbf{c}(x) T(x, t) dx \quad (1e)$$

where  $T(\cdot, t) \in \mathcal{L}^2([0, 1])$  denotes the state variable,  $x \in [0, 1]$  denotes the space variable and  $t \in [0, \infty)$  is the time variable,  $d(x, t)$  is the *unknown* disturbance (including process noise and model uncertainty) in the system,  $f(x, t)$  is

the *unknown* fault,  $\mathbf{y}(t) \triangleq [y_1(t) \ y_2(t) \ \cdots \ y_n(t)]^T \in \mathbb{R}^n$  denotes the  $n$ -dimensional sensor output, and  $\mathbf{c}(x) \triangleq [c_1(x) \ c_2(x) \ \cdots \ c_n(x)]^T \in \mathbb{R}^n$  characterizes the measurement scheme. Without loss of generality, the most frequently used *pointwise measurement* [34], [35] is selected with

$$c_i(x) = \delta(x - \bar{x}_i), \quad i \in \mathcal{N} \triangleq \{1, 2, \dots, n\}$$

corresponding to the pointwise measurement located on  $\bar{x}_i$ ,  $i \in \mathcal{N}$  in the spatial domain  $(0, 1)$  and  $\delta(\cdot)$  denotes the Dirac delta function.

*Assumption 1:* The distributed disturbance  $d(x, t)$  satisfies that

$$\|d(\cdot, t)\|_2 \leq \bar{d} < \infty$$

where  $\bar{d} > 0$  is a *known* constant.

*Remark 1:* It is worth noting that both fault  $f(x, t)$  and disturbance  $d(x, t)$  exist in the system. It is a common setting in the literature that the fault and disturbance exist in both ODE systems [20] and PDE systems [13], [30], [36]. In fact, the disturbance (unknown input)  $d(x, t)$  consists of model uncertainty and process noises. As for the term  $f(x, t)$ , it represents all possible faults and will be 0 in the fault-free case. This setting is essential for robust fault detection; otherwise, a false alarm may arise.

### B. Problem Statement

The problem considered can be formulated as follows:

Use pointwise measurement  $\mathbf{y}(t)$  to design an FDF and generate the corresponding residual signal for *reliable fault detection* for distributed parameter processes that subject to the state-space description of the form in (1a)–(1e).

## III. METHODOLOGY FRAMEWORK

### A. Motivation

The core of model-based fault diagnosis methods for ODE systems lies in the construction of the analytical redundancy [20], which can be used for the residual generation. Bearing this in mind, we intend to construct an FDF in a mimic way.

Note that for ODE systems, both the state variable and output are of finite dimension. However, for systems described by (1), the state variable  $T(x, t)$  is of *infinite dimension* while the output  $\mathbf{y}(t)$  is of finite dimension. One major challenge in designing the FDF is how to map the finite-dimensional output to the infinite-dimensional state variable to complete the so-called *output injection* procedure.

Despite the fact that the backstepping-based boundary observer design shows its superiority in simple design procedures (see [5], [13], [32], [33], and [37] for reference), how to expand it to cover in-domain (pointwise and piecewise [34]) measurement requires more research efforts. This becomes extremely essential for fault detection of distributed parameter processes since in-domain measurement increases the redundancy of the FDF.

To this end, we intend to investigate the FDF design for distributed parameter processes covering both the boundary measurement, pointwise measurement, and piecewise

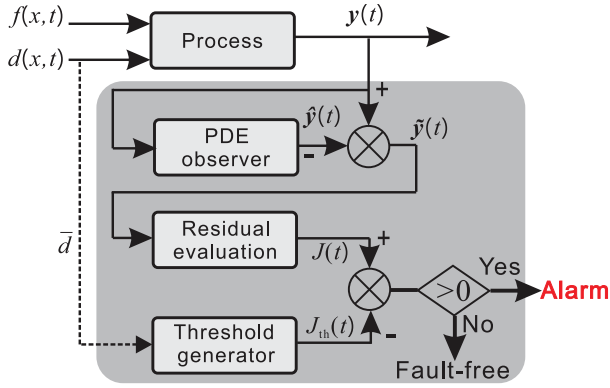


Fig. 1. Methodology framework.

measurement cases motivated by the *unified* Lyapunov-based compensator design [38]. To facilitate the FDF design, some results in [38] are further extended from the spatial  $\mathcal{L}^2$  norm to spatial  $\mathcal{L}^\infty$  norm.

### B. Proposed Framework

Motivated by the above considerations, a framework that consists of an FDF and the corresponding residual evaluation and threshold computation procedures is introduced in Fig. 1 as follows.

## IV. SPATIAL DOMAIN DECOMPOSITION-BASED PDE OBSERVER DESIGN & ANALYSIS

Motivated by the spatial-domain decomposition technique introduced in [34] and [38], the following PDE observer is initiated for system (1a)–(1e):

$$\hat{T}_t(x, t) = \hat{T}_{xx}(x, t) + \mathbf{g}^T(x)\mathbf{L}(\mathbf{y}(t) - \hat{\mathbf{y}}(t)) \quad (2a)$$

$$\hat{T}_x(0, t) = \eta_1 \hat{T}(0, t) \quad (2b)$$

$$\hat{T}_x(1, t) = -\eta_2 \hat{T}(1, t) \quad (2c)$$

$$\hat{\mathbf{y}}(t) = \int_0^1 \mathbf{c}(x)\hat{T}(x, t)dx \quad (2d)$$

where  $\mathbf{g}(x) \triangleq [g_1(x) \ g_2(x) \ \dots \ g_n(x)]^T$  and the elements are defined as

$$g_i(x) \triangleq \begin{cases} 1, & x \in [x_i, x_{i+1}] \\ 0, & x \notin [x_i, x_{i+1}] \end{cases} \quad i \in \mathcal{N} \quad (3)$$

such that  $\bar{x}_i \in (x_i, x_{i+1})$ ,  $i \in \mathcal{N}$ ,  $0 = x_1 < x_2 < \dots < x_n < x_{n+1} = 1$ , and  $\mathbf{L} \triangleq \text{diag}\{l_1, l_2, \dots, l_n\}$  is the gain matrix for the observer. This is the main idea of the spatial-domain decomposition approach [34], [39], as illustrated in Fig. 2.

Introducing the state error variable  $e(x, t) \triangleq T(x, t) - \hat{T}(x, t)$ , and the output error variable  $\tilde{\mathbf{y}}(t) \triangleq \mathbf{y}(t) - \hat{\mathbf{y}}(t)$ , the following error system is obtained:

$$e_t(x, t) = e_{xx}(x, t) + f(x, t) + d(x, t) - \mathbf{g}^T(x)\mathbf{L} \int_0^1 \mathbf{c}(x)e(x, t)dx \quad (4a)$$

$$e_x(0, t) = \eta_1 e(0, t) \quad (4b)$$

$$e_x(1, t) = -\eta_2 e(1, t) \quad (4c)$$

by combining (1) with (2).

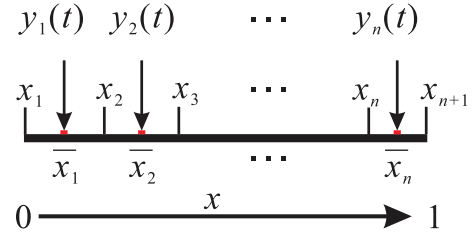


Fig. 2. Illustration of the spatial decomposition approach with pointwise measurement.

*Lemma 1* [38]: Given a scalar function  $z \in \mathcal{H}^1(0, 1)$ , the following inequality holds:

$$\int_0^1 (z(y) - z(\bar{x}))^2 ds \leq 4\phi\pi^{-2} \int_0^1 (dz(y)/dy)^2 dy$$

where  $\bar{x} \in [0, 1]$ ,  $\phi \triangleq \max\{\bar{x}^2, (1 - \bar{x})^2\}$ .

*Lemma 2* [40]: Given regular real functions  $V(t)$  and  $h(t)$

$$\dot{V}(t) \leq -\beta V(t) + h(t) \quad \forall t \geq 0$$

implies that

$$V(t) \leq e^{-\alpha t} V(0) + \int_0^t e^{-\beta(t-v)} h(v) dv \quad \forall t \geq 0$$

for any finite positive constant  $\alpha$ .

*Lemma 3 (Agmon's Inequality [32]):* Given a scalar function  $z \in \mathcal{H}^1(0, 1)$ , the following inequalities hold:

$$\|z(\cdot, t)\|_\infty^2 \leq z^2(0, t) + 2\|z(\cdot, t)\|_2 \|z_y(\cdot, t)\|_2,$$

$$\|z(\cdot, t)\|_\infty^2 \leq z^2(1, t) + 2\|z(\cdot, t)\|_2 \|z_y(\cdot, t)\|_2.$$

*Theorem 1 (Fault Detectability):* For positive constant  $\gamma$ , if there exist positive scalars  $p, q \in (0, (8/\pi^2))$  and  $\bar{l}_i, i \in \mathcal{N}$  such that the following LMIs can be satisfied:

$$\Psi_i + \frac{p\eta_1 q}{2} \mathbf{I} \leq 0, i \in \mathcal{N} \quad (5)$$

where

$$\Psi_i \triangleq \begin{bmatrix} -\frac{p\pi^2}{4\phi_i} & \star & \star & \star \\ \left(\frac{p\pi^2}{4\phi_i} - \frac{\bar{l}_i}{2}\right) & -\frac{p\pi^2}{4\phi_i} & \star & \star \\ 0 & \frac{\bar{l}_i}{2} & -p & \star \\ \frac{p}{2} & 0 & -\frac{p}{2} & -\gamma^2 \end{bmatrix}, \quad i \in \mathcal{N}$$

and  $\phi_i \triangleq \max\{(\bar{x}_i - x_i)^2, (x_{i+1} - \bar{x}_i)^2\}$ , then  $e(x, t)$  is uniformly ultimately bounded (UUB) in the sense of  $\|\cdot\|_\infty$  with ultimate bound  $\bar{e}$  where

$$\bar{e} \triangleq \sqrt{\frac{2}{p \min\{\eta_1, 1\} \bar{\rho}}} \gamma \bar{d} \quad (6)$$

with

$$\bar{\rho} \triangleq \min\left\{\eta_1 q, 2 - \frac{q\pi^2}{4}\right\}. \quad (7)$$

The elements  $l_i$  of the observer gain matrix can be calculated according to

$$l_i = p^{-1} \bar{l}_i, i \in \mathcal{N}. \quad (8)$$

*Proof:* Consider the following Lyapunov function:

$$V(t) \triangleq V_1(t) + V_2(t) \quad (9)$$

where

$$\begin{aligned} V_1(t) &\triangleq \frac{p}{2} \int_0^1 \left( e^2(x, t) + e_x^2(x, t) \right) dx \\ V_2(t) &\triangleq \frac{p\eta_1}{2} e^2(0, t) + \frac{p\eta_2}{2} e^2(1, t). \end{aligned} \quad (10)$$

The time derivative of  $V_1(t)$  can be calculated as

$$\begin{aligned} \dot{V}_1(t) &= \int_0^1 pe(x, t)e_t(x, t)dx \\ &\quad + \int_0^1 pe_x(x, t)e_{xt}(x, t)dx. \end{aligned} \quad (11)$$

Set

$$\bar{l}_i = pl_i, i \in \mathcal{N}.$$

Using integration by parts, one gets

$$\begin{aligned} \int_0^1 pe(x, t)e_t(x, t)dx &= \underbrace{\int_0^1 pe(x, t)e_{xx}(x, t)dx}_{\text{Intergration by parts+(4b)-(4c)}} \\ &\quad - \int_0^1 pe(x, t)g^T(x)Ldx \int_0^1 c(x)e(x, t)dx \\ &\quad + \int_0^1 pe(x, t)d(x, t)dx \\ &= -p\eta_2 e^2(1, t) - p\eta_1 e^2(0, t) \\ &\quad - \int_0^1 pe_x^2(x, t)dx \\ &\quad - \sum_{i=1}^n \int_{x_i}^{x_{i+1}} \bar{l}_i e(x, t)e(\bar{x}_i, t)dx \\ &\quad + \int_0^1 pe(x, t)d(x, t)dx \end{aligned} \quad (12)$$

by combining with (4a) and the boundary conditions in (4b) and (4c).

Moreover, the last term in the RHS of (11) is further formulated as

$$\begin{aligned} &\underbrace{\int_0^1 pe_x(x, t)e_{xt}(x, t)dx}_{\text{Intergration by parts+(4b)-(4c)}} \\ &= -p\eta_2 e(1, t)e_t(1, t) - p\eta_1 e(0, t)e_t(0, t) \\ &\quad - \int_0^1 pe_{xx}(x, t)e_t(x, t)dx \\ &= -p\eta_2 e(1, t)e_t(1, t) - p\eta_1 e(0, t)e_t(0, t) \\ &\quad - \int_0^1 pe_{xx}^2(x, t)dx + \sum_{i=1}^n \int_{x_i}^{x_{i+1}} \bar{l}_i e_{xx}(x, t)e(\bar{x}_i, t)dx \\ &\quad - \int_0^1 pe_{xx}(x, t)d(x, t)dx \end{aligned} \quad (13)$$

considering (4a) and the boundary conditions in (4b) and (4c).

Combining (12) with (13), it can be obtained that

$$\begin{aligned} \dot{V}_1(t) &= -p\eta_2 e^2(1, t) - p\eta_1 e^2(0, t) - p\eta_2 e(1, t)e_t(1, t) \\ &\quad - p\eta_1 e(0, t)e_t(0, t) - \int_0^1 pe_x^2(x, t)dx \\ &\quad - \int_0^1 pe_{xx}^2(x, t)dx - \sum_{i=1}^n \int_{x_i}^{x_{i+1}} \bar{l}_i e(x, t)e(\bar{x}_i, t)dx \\ &\quad + \sum_{i=1}^n \int_{x_i}^{x_{i+1}} \bar{l}_i e_{xx}(x, t)e(\bar{x}_i, t)dx \\ &\quad + \int_0^1 pe(x, t)d(x, t)dx - \int_0^1 pe_{xx}(x, t)d(x, t)dx. \end{aligned} \quad (14)$$

Moreover, it can be calculated that

$$\dot{V}_2(t) = p\eta_1 e(0, t)e_t(0, t) + p\eta_2 e(1, t)e_t(1, t). \quad (15)$$

Combining (14) and (15), it can be derived that

$$\begin{aligned} \dot{V}(t) &= -p\eta_2 e^2(1, t) - p\eta_1 e^2(0, t) - \int_0^1 pe_x^2(x, t)dx \\ &\quad - \int_0^1 pe_{xx}^2(x, t)dx - \sum_{i=1}^n \int_{x_i}^{x_{i+1}} \bar{l}_i e(x, t)e(\bar{x}_i, t)dx \\ &\quad + \sum_{i=1}^n \int_{x_i}^{x_{i+1}} \bar{l}_i e_{xx}(x, t)e(\bar{x}_i, t)dx \\ &\quad + \int_0^1 pe(x, t)d(x, t)dx - \int_0^1 pe_{xx}(x, t)d(x, t)dx. \end{aligned} \quad (16)$$

Using Lemma 1 for each interval  $[x_i, x_{i+1}]$ ,  $i \in \mathcal{N}$ , it can be derived that

$$\int_{x_i}^{x_{i+1}} e_x^2(x, t)dx \geq \frac{\pi^2}{4\phi_i} \int_{x_i}^{x_{i+1}} (e(x, t) - e(\bar{x}_i, t))^2 dx. \quad (17)$$

Substituting (17) into (16) and recalling that  $0 = x_1 < x_2 < \dots < x_n < x_{n+1} = 1$  and  $\eta_1 > 0$ ,  $\eta_2 > 0$ , we have that

$$\begin{aligned} \dot{V}(t) &\leq -p\eta_2 e^2(1, t) - p\eta_1 e^2(0, t) - \sum_{i=1}^n \int_{x_i}^{x_{i+1}} \frac{p\pi^2}{4\phi_i} e^2(x, t)dx \\ &\quad - \sum_{i=1}^n \int_{x_i}^{x_{i+1}} \frac{p\pi^2}{4\phi_i} e^2(\bar{x}_i, t)dx - \sum_{i=1}^n \int_{x_i}^{x_{i+1}} pe_{xx}^2(x, t)dx \\ &\quad + \sum_{i=1}^n \int_{x_i}^{x_{i+1}} \left( \frac{p\pi^2}{2\phi_i} - \bar{l}_i \right) e(x, t)e(\bar{x}_i, t)dx \\ &\quad + \sum_{i=1}^n \int_{x_i}^{x_{i+1}} \bar{l}_i e_{xx}(x, t)e(\bar{x}_i, t)dx \\ &\quad + \sum_{i=1}^n \int_{x_i}^{x_{i+1}} pe(x, t)d(x, t)dx \\ &\quad - \sum_{i=1}^n \int_{x_i}^{x_{i+1}} pe_{xx}(x, t)d(x, t)dx \\ &= \sum_{i=1}^n \int_{x_i}^{x_{i+1}} \xi_i^T(x, t)\Psi_i \xi_i(x, t)dx - p\eta_2 e^2(1, t) \\ &\quad - p\eta_1 e^2(0, t) + \gamma^2 \|d(\cdot, t)\|_2^2 \end{aligned} \quad (18)$$

where

$$\xi_i(x, t) \triangleq [e(x, t) \quad e(\bar{x}_i, t) \quad e_{xx}(x, t) \quad d(x, t)]^T.$$

By substituting (5) into (18) and considering Assumption 1, it can be obtained that

$$\begin{aligned} \dot{V}(t) &\leq -\frac{p\eta_1 q}{2} \sum_{i=1}^n \int_{x_i}^{x_{i+1}} \xi_i^T(x, t) \Psi_i \xi_i(x, t) dx \\ &\quad - p\eta_2 e^2(1, t) - p\eta_1 e^2(0, t) + \gamma^2 \bar{d}^2 \\ &\leq -\frac{p\eta_1 q}{2} \left( \|e(\cdot, t)\|_2^2 + \|e_{xx}(\cdot, t)\|_2^2 \right) - p\eta_2 e^2(1, t) \\ &\quad - p\eta_1 e^2(0, t) + \gamma^2 \bar{d}^2 \\ &\leq -\left( \eta_1 q \frac{p}{2} \|e(\cdot, t)\|_2^2 + \frac{\pi^2 \eta_1 q p}{4} \frac{p}{2} \|e_{xx}(\cdot, t)\|_2^2 \right) \\ &\quad + 2 \frac{p\eta_2}{2} e^2(1, t) + \left( 2 - \frac{q\pi^2}{4} \right) \frac{p\eta_1}{2} e^2(0, t) + \gamma^2 \bar{d}^2 \\ &\leq -\bar{\rho} V(t) + \gamma^2 \bar{d}^2 \end{aligned} \quad (19)$$

by applying Lemma 1 and recalling the definition of  $\bar{\rho}$  in (7).

Hence, one gets

$$V(t) \leq V(0) \exp(-\bar{\rho}t) + \frac{\gamma^2 \bar{d}^2}{\bar{\rho}} \quad (20)$$

by Lemma 2.

Recalling Lemma 3 and Young's inequality [32], we get

$$\begin{aligned} \|e(\cdot, t)\|_\infty^2 &\leq e^2(0, t) + 2\|e(\cdot, t)\|_2 \|e_x(\cdot, t)\|_2 \\ &\leq e^2(0, t) + \|e(\cdot, t)\|_2^2 + \|e_x(\cdot, t)\|_2^2 \\ &\leq \bar{e}_B^2(t) \triangleq \frac{2V(0)}{p \min\{\eta_1, 1\}} \exp(-\bar{\rho}t) + \bar{e}^2 \end{aligned} \quad (21)$$

using (20).

Recalling Assumption 1, one gets that  $0 < \bar{e} < \infty$ . Hence,  $e(x, t)$  satisfies the UUB condition in the sense of  $\|\cdot\|_\infty$  while the ultimate bound is  $\bar{e}$ . ■

*Remark 2:* Note that when using the Poincaré–Wirtinger inequality's variants and Agmon's inequality for deriving (19) and (21), respectively,  $e(0, t)$  is considered in both inequalities. One can consider replacing it with  $e(1, t)$  instead, and LMIs in (5) change to

$$\Psi_i + \frac{p\eta_2 q}{2} \mathbf{I} \leq 0, i \in \mathcal{N}$$

corresponding under this condition. Moreover, one can get an ultimate bound following the same procedures. The details are not discussed due to triviality.

*Remark 3:* The proposed spatial domain decomposition approach contains the boundary measurement at  $x = 0$  or  $x = 1$  if  $\bar{x}_1 = 0$  for  $\bar{x}_{n+1} = 1$ . Moreover, the results can be easily expanded to the piecewise measurement case, as shown in [38]. Furthermore, the results are also applicable for the homogeneous Dirichlet boundary conditions, homogeneous Neumann boundary conditions, mixed homogeneous Neumann–Dirichlet boundary conditions, and mixed homogeneous Dirichlet–Neumann boundary conditions. These extensions are trivial exercises and are not discussed.

## V. FDF DESIGN & ANALYSIS

For the fault detection purpose, the following residual signal  $\mathbf{r}(t)$  is selected intuitively:

$$\mathbf{r}(t) = \tilde{\mathbf{y}}(t). \quad (22)$$

Then, we show the *lumped* evaluation scheme and the threshold as follows:

$$J(t) = \int_t^{t+\Delta} |\mathbf{r}(\tau)|_2^2 d\tau \quad (23a)$$

$$J_{\text{th}}(t) = \sup_{f(x,t)=0} J(t) \quad (23b)$$

where  $\Delta > 0$  is the length of the evaluation time window.

*Remark 4:* There are several residual evaluation functions that are widely used in fault detection, such as the peak value, the average value, the root-mean-square (RMS) value, and the  $\mathcal{L}_2$  norm [20]. The RMS value and  $\mathcal{L}_2$  norm are related and are designed to reduce the false alarm rate (FAR) at the cost of fault detectability. It is worth noticing that  $(\int_t^{t+T} |\mathbf{r}(\tau)|_2^2 d\tau)^{1/2}$  is a general form of the  $\mathcal{L}_2$  norm [20], [41] of the signal  $\mathbf{r}(t)$ , which measures the energy of a signal over a time interval  $[t, t+T]$ . Since evaluation over the whole time domain is usually unrealistic, introducing an evaluation window is a practical modification [20]. This kind of evaluation function was widely used in the literature discussing FD issues of dynamic systems (see [42] and [43]).

Hence the following decision logic will guarantee reliable detection performance:

$$\begin{cases} J(t) > J_{\text{th}}(t) \implies \text{faulty at time } t \\ \text{Otherwise} \implies \text{fault-free at time } t. \end{cases} \quad (24)$$

*Theorem 2:* (Fault Detection Threshold Calculation) Under the lumped residual evaluation function in (23), if the conditions in Theorem 1 are satisfied, the time-varying detection threshold  $J_{\text{th}}(t)$  for fault detection can be computed as

$$J_{\text{th}}(t) = \int_t^{t+\Delta} n e_B^2(\tau) d\tau. \quad (25)$$

*Proof:* The proof is based on (21) and the property of  $\|\cdot\|_\infty$ . It is omitted due to triviality.

*Remark 5:* It is worth noting that the threshold  $J_{\text{th}}(t)$  is designed to be time varying rather than a constant that is frequently used [29], [33]. The intuitive idea is that the initial error  $e_0(x)$  contributes to  $V(0)$  and the time-varying term in (25) is decreasing exponentially with time, if only the ultimate bound  $\bar{e}$  is considered in the threshold setting, one inevitable outcome is the false alarm in the initial transition stage [29]. To decrease the FAR, we add the compensate time-varying term in the threshold.

## VI. ILLUSTRATIVE EXAMPLE

### A. System Description

Consider the following thermal model of a hot strip mill [24], [44]–[46] under the *fault-free* and *disturbance-free* conditions:

$$\rho c \bar{T}_\tau(\xi, \tau) = k \bar{T}_{\xi\xi}(\xi, \tau) \quad (26a)$$

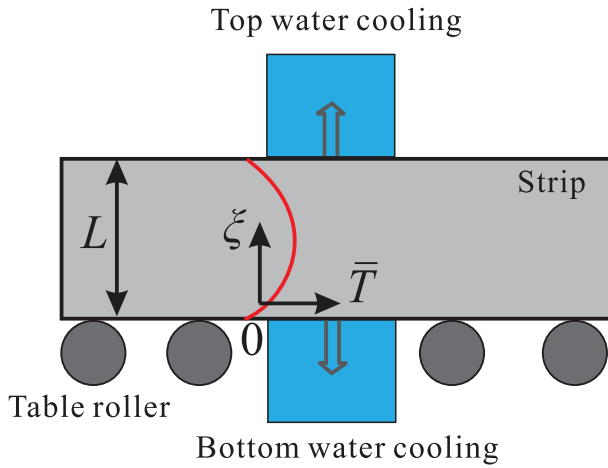


Fig. 3. Hot strip mill cooling system.

TABLE I  
PROCESS PARAMETER SETTING

	Definition (Unit)	Value
$L$	Thickness ([m])	0.01
$\rho$	Density ([Kg/m <sup>3</sup> ])	7850
$k$	Average thermal conductivity ([W/m ° C])	28
$c$	Average specific heat ([J/Kg ° C])	932
$h$	Heat transfer coefficients([W/m <sup>2</sup> ° C])	2000
$T_w$	Temperature of the cooling water (l° C)	30

$$\bar{T}_\xi(0, \tau) = \frac{h}{k}(\bar{T}(0, \tau) - T_w) \quad (26b)$$

$$\bar{T}_\xi(L, \tau) = -\frac{h}{k}(\bar{T}(L, \tau) - T_w) \quad (26c)$$

$$\bar{T}(\xi, 0) = \bar{T}_0(\xi) \quad (26d)$$

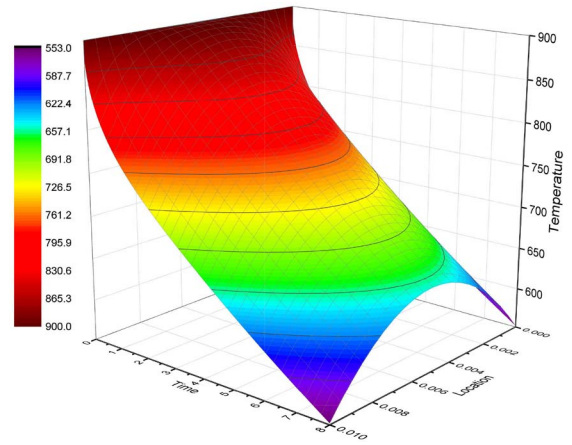
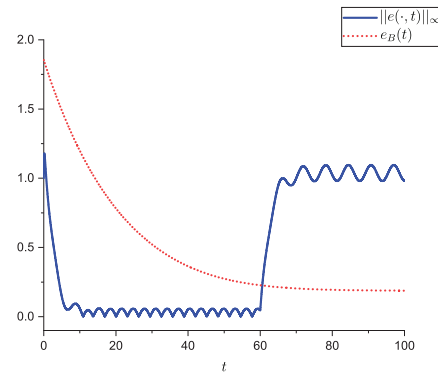
where  $\bar{T}(\xi, \tau) \in \mathcal{L}^2([0, L])$  denotes the temperature distribution along the thickness direction and  $L$  is the thickness of the strip,  $\xi \in [0, L]$  is the space variable,  $\tau \in [0, \infty)$  denotes the time,  $\rho$  is the density,  $k$  denotes the average value of thermal conductivity while  $c$  is the average value of specific heat,  $h$  denotes the heat transfer coefficient, and  $T_w$  denotes the temperature of water. Fig. 3 presents the layout of a strip on the ROT.

To facilitate the analysis, considering the disturbance and fault existing in the process and applying the scaling transformation [32], [44]

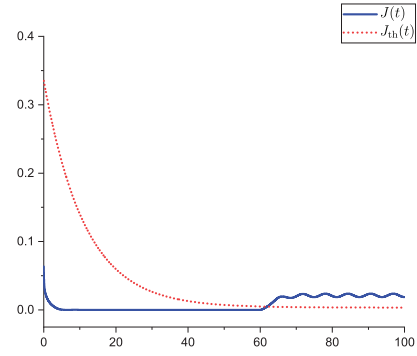
$$\begin{aligned} x &= \frac{\xi}{L} \\ t &= \frac{k\tau}{\rho c L^2} \\ T &= \frac{\bar{T} - T_w}{\bar{T}_0 - T_w} \end{aligned}$$

to system (26) leads to the following normalized equation:

$$\begin{aligned} T_t(x, t) &= T_{xx}(x, t) + d(x, t) + f(x, t) \\ T_x(0, t) &= \eta T(0, t) \\ T_x(1, t) &= -\eta T(1, t) \\ T(x, 0) &= T_0(x) = 1 \\ y(t) &= T(0, t) \end{aligned}$$

Fig. 4. Temperature distribution profile  $\bar{T}(\xi, \tau)$  under normal operating condition.

(a)



(b)

Fig. 5. Simulation results with abrupt fault. (a) Trajectory of  $\|e(\cdot, t)\|_\infty$  and  $e_B(t)$ . (b) Trajectory of  $J(t)$  and  $J_{th}(t)$ .

where  $\eta \triangleq \frac{hL}{k} > 0$ . For practical consideration, the *boundary* measurement  $T(0, t)$  on the bottom of the strip is used for fault detection.

### B. Parameter Setting

Typically, the strip's temperature decreases from nearly 900 °C to 650 °C [44]; hence, the initial condition in (26d) is selected as  $\bar{T}_0(\xi) = 900$  without loss of generality. Moreover, as illustrated in [44] and [47], the parameters used in the

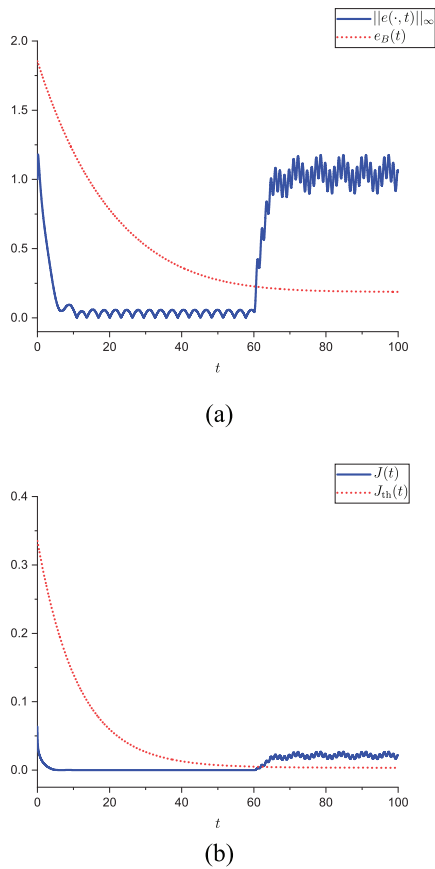


Fig. 6. Simulation results with periodic fault. (a) Trajectory of  $\|e(\cdot, t)\|_\infty$  and  $e_B(t)$ . (b) Trajectory of  $J(t)$  and  $J_{th}(t)$ .

simulation along with their practical values are summarized in Table I.

### C. Simulation Results

The simulation result of system (26) under normal operating condition is presented in Fig. 4. The following disturbance is considered:

$$d(x, t) = 0.05 \exp\left(-0.5(x - 0.5)^2\right) \sin(t).$$

Furthermore, the following fault is injected:

$$f(x, t) = b_f(x)f(t)$$

where

$$b_f(x) = H(x - 0.5) - H(x - 0.75)$$

is the shape function of the fault with  $H(\cdot)$  denoting the standard Heaviside function. Two kinds of faults are considered

$$\text{Abrupt fault: } f(t) = \begin{cases} 0, & t \in [0, 60) \\ 1, & t \in [60, \infty) \end{cases}$$

$$\text{Periodic fault: } f(t) = \begin{cases} 0, & t \in [0, 60) \\ 1 + \sin(5t), & t \in [60, \infty). \end{cases}$$

Selecting the evaluation time window  $\Delta = 0.1$ , the simulation results of both abnormal fault and periodic fault are presented in Figs. 5 and 6, respectively. One can find that the fault detection time  $t = 60$  can be well detected for both faults.

Moreover, results in Figs. 5(a) and 6(a) show that  $\|e(\cdot, t)\|_\infty$  is UUB in the fault-free scenario, which demonstrated the correctness of Theorem 1.

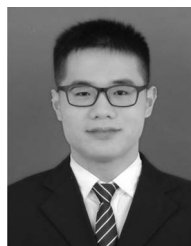
## VII. CONCLUSION

In this article, a unified framework for fault detection consisting of both the boundary and in-domain sensors was proposed without modal approximation or interpolation. An FDF based on the PDE observer was developed, which is motivated by the FDF design for LPSs. Rigorous theoretic analysis was presented to guarantee the reliability of the proposed detection scheme. Motivated by the fact that for real industrial processes, the system model may not be completely known, hence it would be interesting to further investigate the fault detection problem of these processes with partially known or even unknown system parameters in the future.

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